

**MODELLING PARALLEL OPERATION OF
POWER RECTIFIERS WITH PSPICE**

B. Rivet

I - INTRODUCTION

The behaviour of semiconductor components is always linked with the junction temperature.

This is the case, for example, in current-sharing between diodes connected in parallel. The current in each diode depends on the forward characteristic but also on the junction temperature.

This study describes thermal and electrical modelling of diodes connected in parallel. It allows the variation in current and junction temperature of each diode to be visualized between turn on and the equilibrium state.

II - ELECTRICAL AND THERMAL MODELLING OF A DIODE

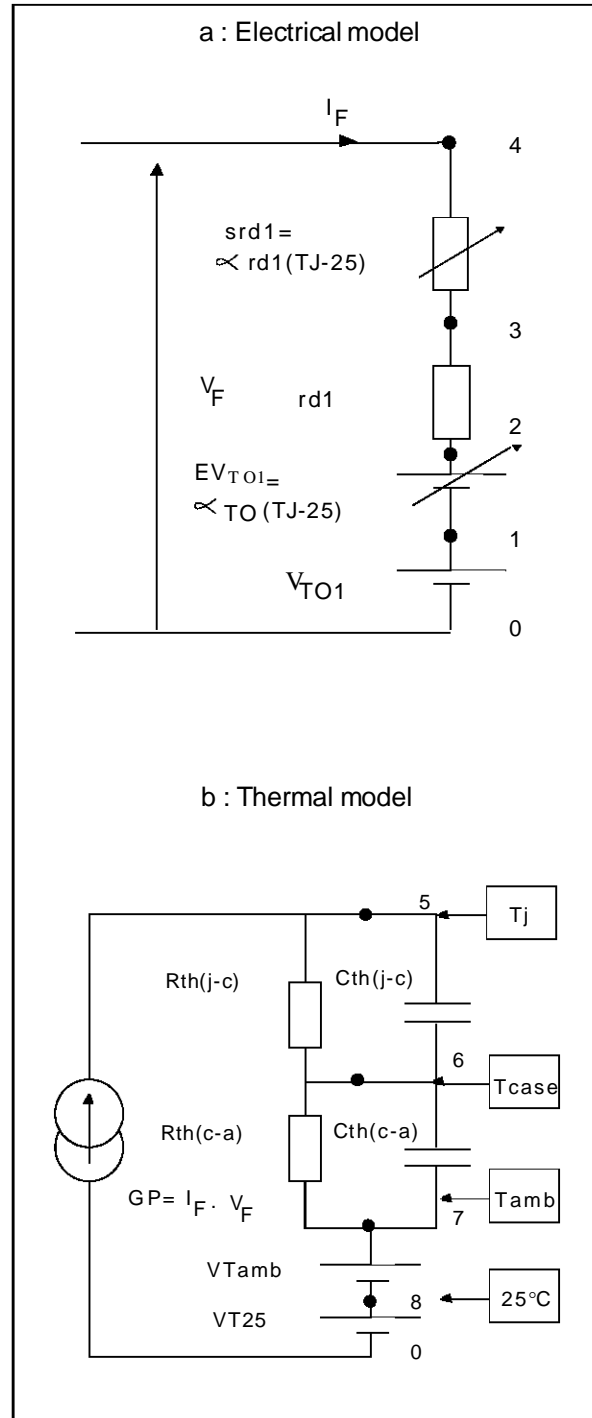
The forward characteristic of a diode is modelled by a threshold voltage V_{T0} in series with a dynamic resistance rd . These two parameters depend on the junction temperature of the diodes. V_{T0} has a negative temperature coefficient $\propto T_0$ and rd have a positive temperature coefficient $\propto rd$.

One way to simulate the operation of such a device is to split the model into 2 different circuits : one "electrical" model and one "thermal" model.

The electrical and thermal models of a diode are shown in fig.1.

The thermal model is represented by electrical components.

Fig.1 : Electrical and thermal models of a diode



APPLICATION NOTE

THE ELECTRICAL PARAMETERS ARE :

- V_{T01} : Threshold voltage at $T_j = 25^\circ\text{C}$
 $E_{V_{T01}}$: Threshold voltage versus junction temperature
 $E_{V_{T01}} = \infty_{T0} (T_j - T_{amb})$
 $= \infty_{T0} (V(5) - V(8))$
 $rd1$: Dynamic resistance at $T_j = 25^\circ\text{C}$
 $srd1$: Dynamic resistance versus temperature
 $srd1 = \infty_{rd1} (T_j - T_{amb})$
 $= \infty_{rd1} (V(5) - V(8))$
 V_F : Instantaneous forward voltage across the diode
 I_F : Instantaneous forward current in the diode

THE THERMAL PARAMETERS ARE :

- GP : Generator current representing the instantaneous power dissipated in the diode.
 $GP = V_F \times I_F$
 $R_{th(j-c)}$: Resistor representing the junction to case thermal resistance
 $C_{th(j-c)}$: Capacitor representing the junction to case thermal capacitance
 $R_{th(c-a)}$: Resistor representing the case to ambient thermal resistance
 $C_{th(c-a)}$: Capacitor representing the case to ambient thermal capacitance
 V_{Tamb} : Voltage generator representing the ambient temperature
 $V_{Tamb} = T_{amb} - 25^\circ\text{C}$
 V_{T25} : Voltage generator representing the 25°C temperature

III - MODELLING OF SEVERAL DIODES IN PARALLEL

See the application note : "Parallel operation of power rectifiers" (B.RIVET) for the qualitative analysis. In this note the acceptable difference between forward voltage drops (ΔV_F) is calculated so that the diodes can be safely connected in parallel.

The modelling will be based on the worst case situation. Suppose D1 has the lowest V_{T0} and rd and the highest $R_{th(j-c)}$ so that this diode supports the highest current. The diode D2 and D3 have the same characteristics (maximum V_{T0} and rd , minimum $R_{th(j-c)}$).

The electrical and thermal models are shown in Fig.2.

Fig.2a : Thermal model of 3 rectifiers in parallel

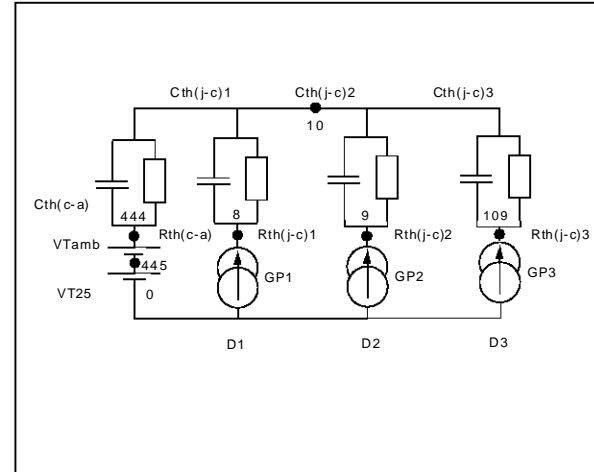
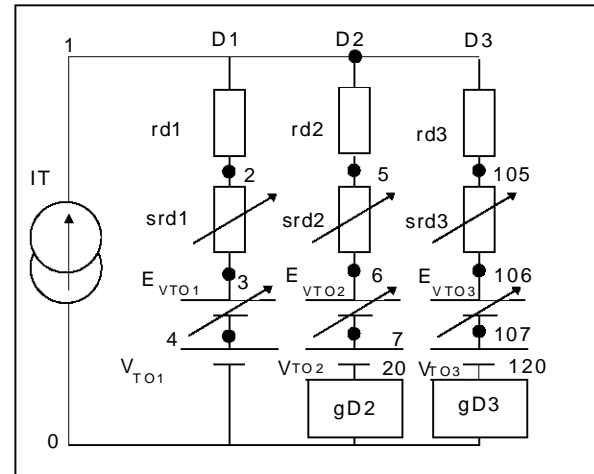


Fig.2b : Electrical model of 3 rectifiers in parallel



The modelling of several diodes in parallel will be treated with an example. Consider three BYV255 in parallel. The total current I_T is rectangular with a duty cycle equal to 0.5, a peak current of 300 A and a frequency of 100 HZ.

In this example the following values have been taken.

- $rd1 = 1.2 \text{ m}\Omega$
 $rd2 = rd3 = 0.9 \text{ m}\Omega$
 $V_{T01} = 0.82 \text{ V}$
 $\infty_{T0} = -1.6 \times 10^{-3} \text{ V}/^\circ\text{C}$
 $\infty_{rd} = 2.10^{-6} \Omega/^\circ\text{C}$
 $R_{th(j-c)1} = 0.4^\circ\text{C}/\text{W}$
 $R_{th(j-c)2} = R_{th(j-c)3} = 0.3^\circ\text{C}/\text{W}$

Their calculation is explained in the previously mentioned application note.

$C_{th(j-c)1}$, $C_{th(j-c)2}$, $C_{th(j-c)3}$, $C_{th(j-c)}$ are not very important parameters; they influence only the

transient behaviour of the circuit and not the equilibrium state.

In the example $C_{th(j-c)1} = C_{th(j-c)2} = C_{th(j-c)3} = C_{th(c-a)} = 1 \text{ sW/}^\circ\text{C}$ has been assumed.

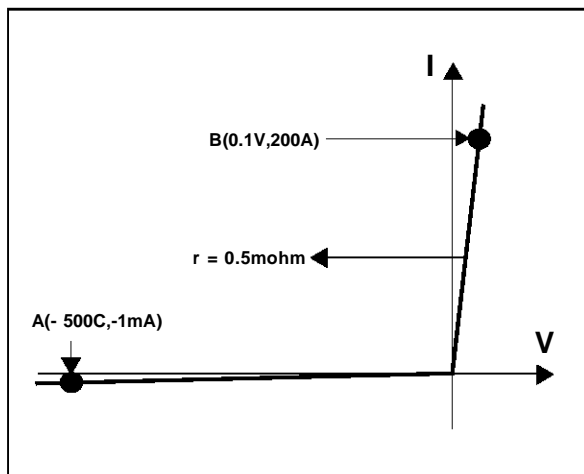
The simulation has been done with $\Delta V_F = 80 \text{ mV}$

Consider

$$\Delta F \approx \Delta V_{T0} \text{ so } V_{T02} = V_{T03} = V_{T01} + \Delta V_F = 0.9 \text{ V}$$

In the electrical modelling of the diode D2 and D3, components gD2, gD3 have been added. Their characteristic is drawn in Fig.3

Fig.3 : Characteristic of gD2 and gD3



These auxiliary components are needed to avoid a current flow in the diodes when the current I_T is equal to zero. When the diodes conduct, these components are equivalent to the resistance of $0.5 \text{ m}\Omega$.

To take these resistances into account, make $rd2 = rd3 = 0.4 \text{ m}\Omega$

The PSPICE description of this circuit is given in the appendix.

Fig.4 and 5 show the results of the simulation.

These curves represent the variations of the current in D1 and D2 (fig.4) and the junction temperatures T_{j1} and T_{j2} (fig.5)

At $t = 0 \text{ s}$ $T_{j1} = T_{j2} = T_{amb} = 40^\circ\text{C}$: the current is higher in D1 than in D2 so T_{j1} will increase more quickly than T_{j2} and the difference between I_{F1} and I_{F2} will increase also.

When the equilibrium state is reached this difference becomes constant.

If the frequency and the thermal capacitor are high the simulation time needed to reach the equilibrium state is long. So to reduce this time the C_{th} values can be decreased (This change will affect only the transient behaviour).

IV - CONCLUSION

A very flexible analysis of operation of several rectifiers in parallel can be done easily by using simulation on PSPICE. Because of the importance of the thermal effect on the various parameters of each diode we have design 2 models operating simultaneously. The first circuit calculates the electrical parameters while the second monitors the junction temperature.

One example of calculation with 3 diodes in parallel has shown how the curves of the currents and the junction temperatures can be obtained.

This double model is very powerful. The designer can add other diodes, insert wiring resistance, change the current waveform and also the cooling characteristic of the design.

APPLICATION NOTE

Fig.4 : Variation of the current in D1 (IRD1) and in D2 (IRD2) versus time

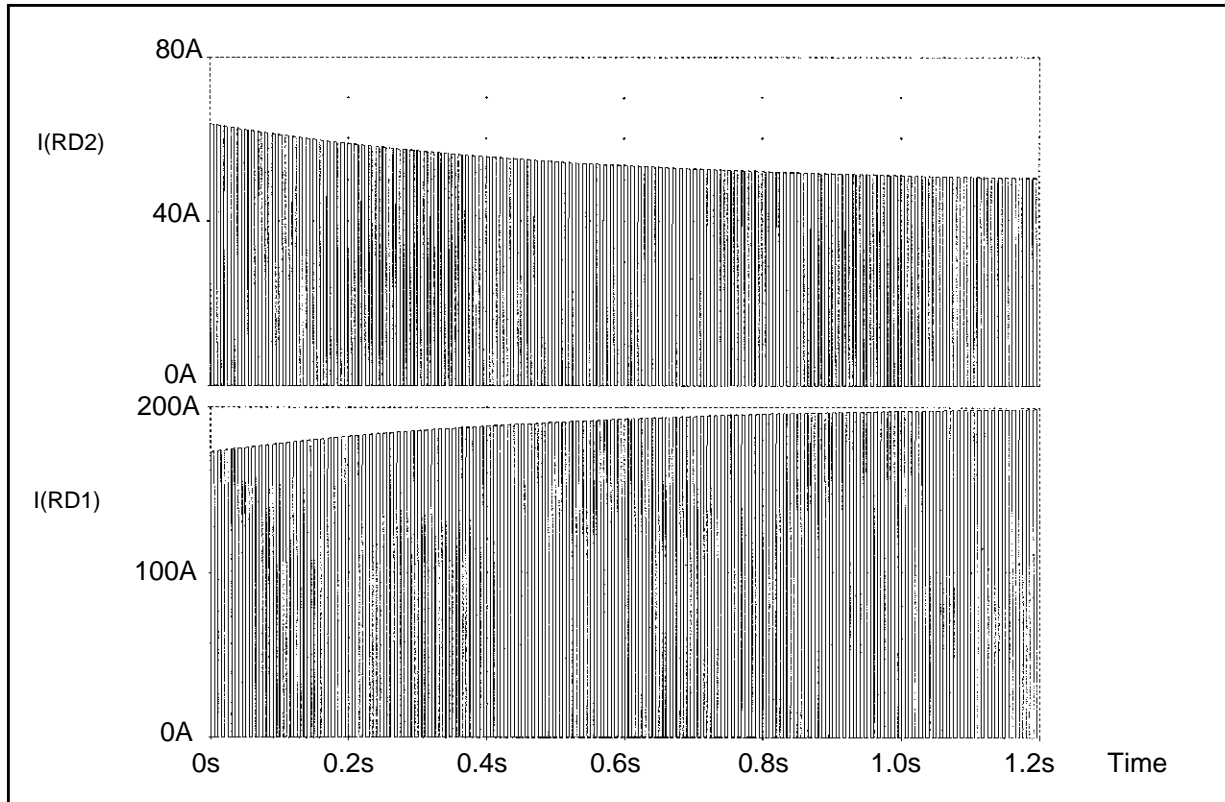
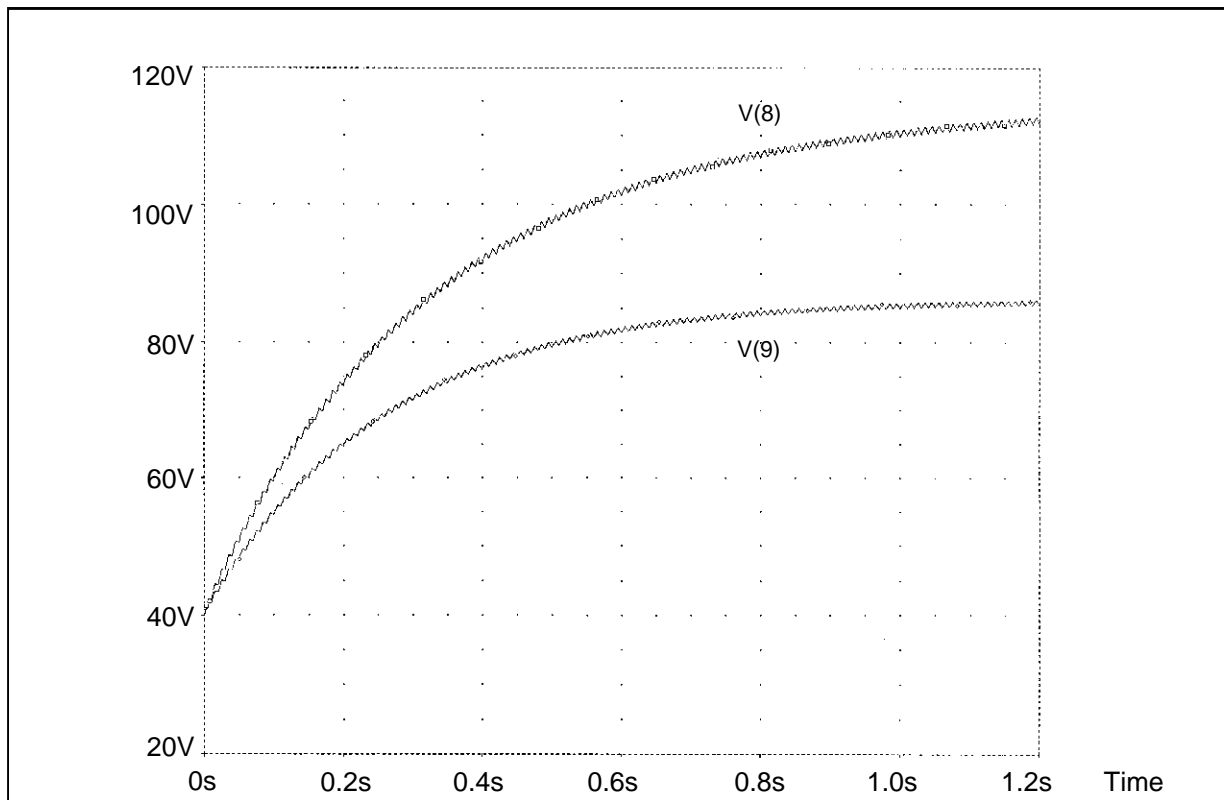


Fig.5 : Variation of the junction temperature of the diode D1 (V(8)) and D2 (V(9)) versus time



APPENDIX

DIODES-PARALLEL

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I1 1 0 PULSE50 -300 0 0 0 5m 10m)
RD1 1 2 .9m
SRD1 2 3 8 445 SMOD
EVTO1 3 4 VALUE={-0.0016*V(8,445)}
VTO1 4 0 0.82
RD2 1 5 0.7m
SRD2 5 6 9 445 SMOD
EVTO2 6 7 VALUE={-0.0016*V(9,445)}
VTO2 7 20 0.9
GD2 20 0 TABLE {V(20)} =
+ (-500,-0.001) (0,0) (0.1,200)
RD3 1 105 0.7m
SRD3 105 106 109 445 SMOD
EVTO3 106 107 VALUE={-0.0016*V(109,445)}
VTO3 107 120 0.9
GR3 120 0 TABLE {V(120)} =
+ (-500,-0.001) (0,0) (0.1,200)
GP1 8 0 VALUE={-V(1,0)*V(1,2)*1111.11}
RTHJC1 8 10 .4
CTHJC1 8 10 1
GP2 9 0 VALUE={-V(1,0)*V(1,5)*1428.57}
RTHJC2 9 10 .3
CTHJC2 9 10 1
GP3 109 0 VALUE={-V(1,0)*V(1,105)*1428.57}
RTHJC3 109 10 .3
CTHJC3 109 10 1
RTHCA 10 444 .3
CTHCA 10 444 1
VTAMB 444 445 15
VT25 445 0 25
MODEL SMOD VSWITCH
(ROFF=0.0006 ROFF=0.00001 VON=250
VOFF=25)
tran 100.000u 1.2 0 0 ;*ipsp*
END

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